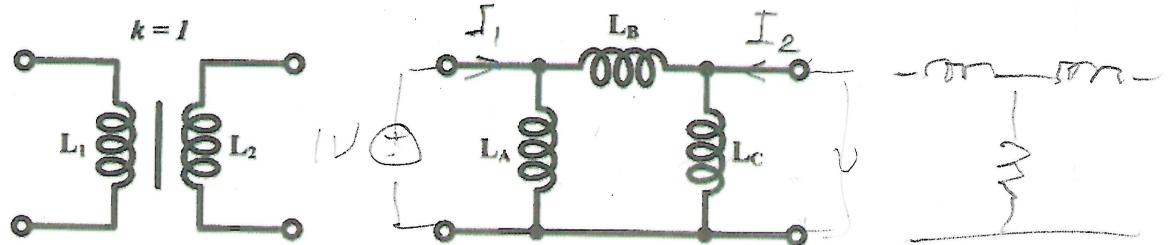


1. Find the values in the "pi"-equivalent shown of the "physical" transformer for general values of  $k$ . What happens to the model if  $k=1$  (close coupling)?

$$M^2 = L_1 L_2$$



$$V_1 = sL_1 I_1 + sM I_2$$

$$V_2 = sM I_1 + sL_2 I_2$$

$$\begin{bmatrix} Y_1 \\ Y_2 \end{bmatrix} = \begin{bmatrix} sL_1 & sM \\ sM & sL_2 \end{bmatrix}$$

$$Y = \frac{1}{s^2(L_1 L_2 - M^2)} \begin{bmatrix} sL_2 & -sM \\ -sM & sL_1 \end{bmatrix},$$

$$Y_{11} = \frac{L_2}{s(L_1 L_2 - M^2)} = \frac{1}{sL_A} + \frac{1}{sL_B}$$

$$Y_{12} = Y_{21} = \frac{-M}{s(L_1 L_2 - M^2)} = -\frac{1}{sL_B}$$

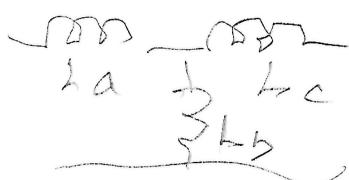
$$Y_{22} = \frac{L_1}{s(L_1 L_2 - M^2)} = \frac{1}{sL_C} + \frac{1}{sL_B}$$

$$\frac{1}{sL_A} = Y_{11} + Y_{12} = \frac{L_2 - M}{s(L_1 L_2 - M^2)}$$

$$L_A = \frac{L_1 L_2 - M^2}{L_2 - M} \rightarrow L_C = \frac{L_1 L_2 - M^2}{L_1 - M}$$

$$L_B = \frac{L_1 L_2 - M^2}{M}, \quad L_1 L_2 = M^2/k$$

For  $k=1$ ,  $L_A = L_B = L_C = 0$ , the model cannot be used. The T-equivalent is OK:



2.

$$V_{in} = I_{in} \left( \frac{1}{r_{in}} + s(C_{in} + C_{fb}) \right)^{-1}$$

$$I_{out} = g_m V_{in} - sC_{fb} V_{in}$$

$$\Rightarrow I_{in}(g_m - sC_{fb}) \left[ \frac{1}{r_{in}} + s(C_{in} + C_{fb}) \right]^{-1}$$

$$A_I = \frac{g_m - j\omega C_{fb}}{\frac{1}{r_{in}} + j\omega(C_{in} + C_{fb})} \quad s = j\omega$$

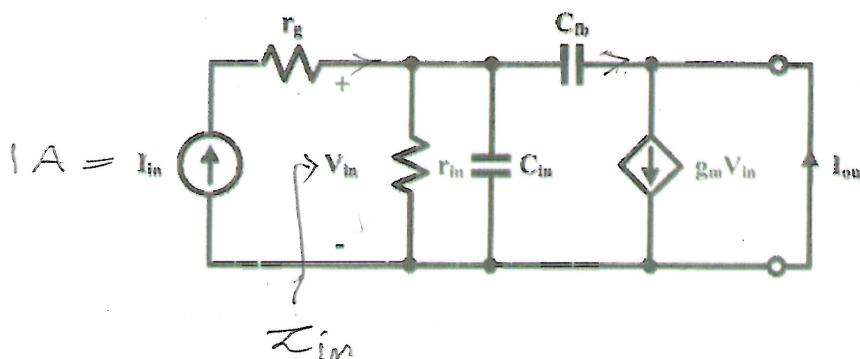
$$|A_I|^2 = \frac{g_m^2 + \omega^2 C_{fb}^2}{\frac{1}{r_{in}^2} + \omega^2(C_{in} + C_{fb})^2} \rightarrow 1$$

$$g_m^2 - \frac{1}{r_{in}^2} = \omega_T^2 [(C_{in} + C_{fb})^2 - C_{fb}^2]$$

$$\omega_T^2 = \left[ \frac{g_m^2 - 1/r_{in}^2}{C_{in}^2 + 2C_{in}C_{fb}} \right]$$

$$f_T = \frac{1}{2\pi} \left[ \frac{g_m^2 - 1/r_{in}^2}{C_{in}^2 + 2C_{in}C_{fb}} \right]^{1/2}$$

2. The circuit shown below is the small-signal model of a semiconductor device. Find  
 a. the short-circuit current gain  $A_I(j\omega) = I_o(j\omega)/I_{in}(j\omega)$ , and b. the unit-gain frequency  $f_T$ , i.e., the frequency where  $|A_I| = 1$ .



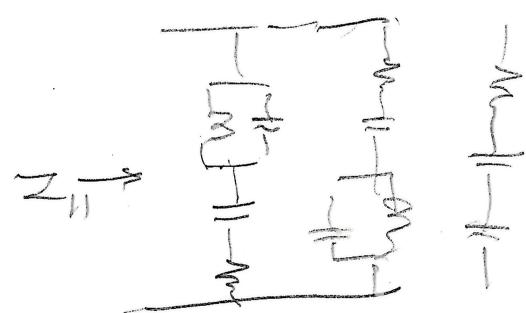
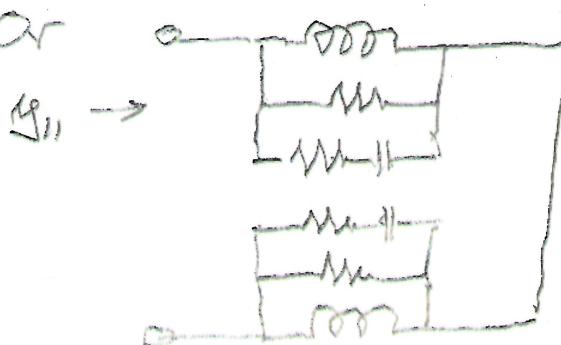
$$3. \quad Y_{11} \Big|_{S \rightarrow \infty} \rightarrow 1/2sL \doteq 1/Z_1$$

$$Z_1 = 2sL$$

$$Z_{11} \Big|_{S \rightarrow \infty} \rightarrow \frac{1}{2} \left( 2R + \frac{1}{sC} \right) = R + \frac{1}{2sC} \doteq R + Z_2$$

$$Z_2 = \frac{1}{2sC}$$

Or

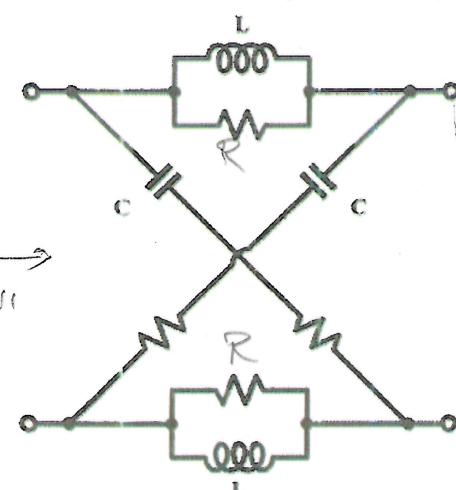


$$Y_{11} = \frac{1}{2} \left[ \frac{1}{sL} + \frac{1}{R} + \frac{1}{R + 1/sC} \right] \doteq \frac{1}{Z_1} + \frac{1}{R + RZ_2/(R+Z_2)}$$

$$S \rightarrow 0; \quad Z_1 = 2sL$$

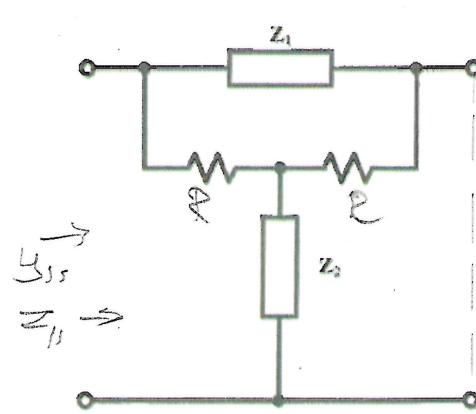
$$\frac{2R + 1/sC}{2R(R + 1/sC)} = \frac{2(R + Z_2)}{2(R^2 + 2RZ_2)} \Rightarrow Z_2 = 1/2sC$$

3. The two two-ports shown below have the same  $Y$ ,  $Z$ , etc. matrices. All resistors have the value  $R$ , where  $R^2 = L/C$ . What are the impedances  $Z_1(s)$  and  $Z_2(s)$ ? *Hint:* consider the behavior for very low and very high frequencies!



Lattice

$$Y_{11}$$



$$Y_{11}$$

Black box

