

ECE 464/564 : HW 2 Solution

1. a)
$$Y(z) \left(1 - \frac{3}{4} z^{-1} + \frac{1}{8} z^{-2} \right) = X(z) \left(1 - \frac{3}{8} z^{-1} \right)$$

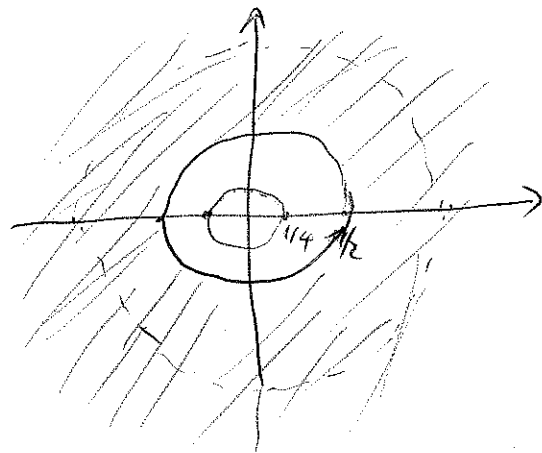
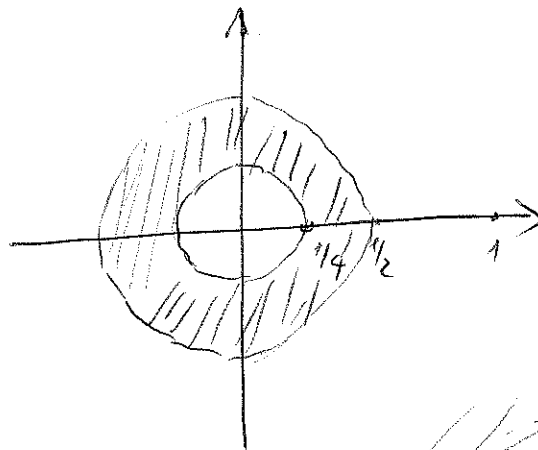
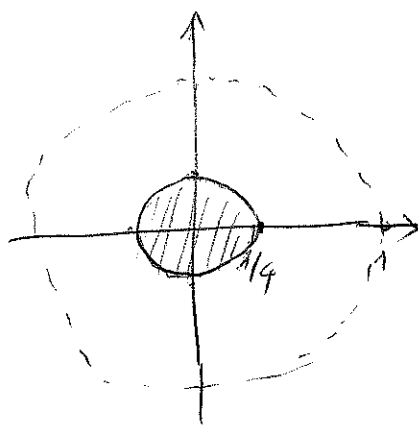
$$\Rightarrow H(z) = \frac{Y(z)}{X(z)} = \frac{1 - \frac{3}{8} z^{-1}}{1 - \frac{3}{4} z^{-1} + \frac{1}{8} z^{-2}}$$

b)
$$H(z) = \frac{1 - \frac{3}{8} z^{-1}}{\left(1 - \frac{1}{2} z^{-1} \right) \left(1 - \frac{1}{4} z^{-1} \right)} = \frac{1/2}{1 - \frac{1}{2} z^{-1}} + \frac{1/2}{1 - \frac{1}{4} z^{-1}}$$

* For $|z| < \frac{1}{4}$: $h[n] = -\frac{1}{2} \left(\frac{1}{2} \right)^n u[-n-1] - \frac{1}{2} \left(\frac{1}{4} \right)^n u[-n-1]$

* For $\frac{1}{4} < |z| < \frac{1}{2}$: $h[n] = \frac{1}{2} \left(\frac{1}{4} \right)^n u[n] - \frac{1}{2} \left(\frac{1}{2} \right)^n u[-n-1]$

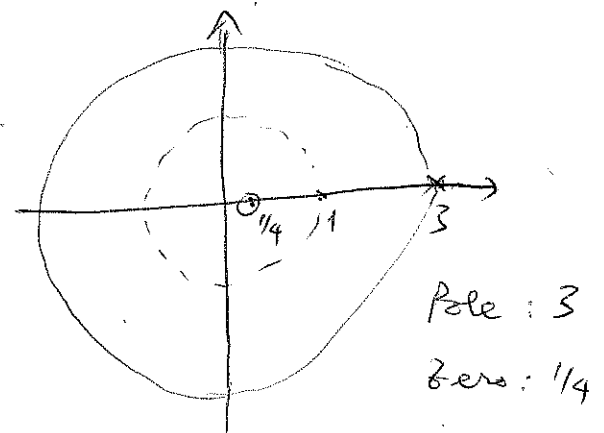
* For $|z| > \frac{1}{2}$: $h[n] = \frac{1}{2} \left(\frac{1}{4} \right)^n u[n] + \frac{1}{2} \left(\frac{1}{2} \right)^n u[n]$



2

$$X_1(z) = \frac{1 - \frac{1}{4}z^{-1}}{1 - 3z^{-1}}$$

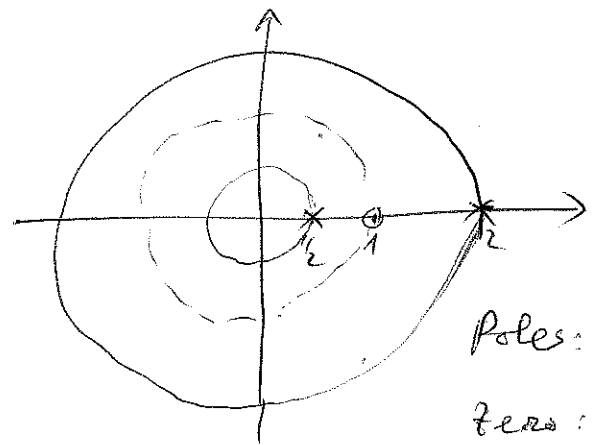
- *) $X_1(z)$ is stable for $|z| < 3$
- *) $X_1(z)$ is causal for $|z| > 3$
- *) There is no ROC where $X_1(z)$ is both stable and causal.



Pole: 3
Zero: 1/4

$$X_2(z) = \frac{1 - z^{-1}}{1 - \frac{5}{2}z^{-1} + z^{-2}}$$

- *) $X_2(z)$ is stable for $\frac{1}{2} < |z| < 2$
- *) $X_2(z)$ is causal for $|z| > 2$
- *) There is no ROC where $X_2(z)$ is both stable and causal

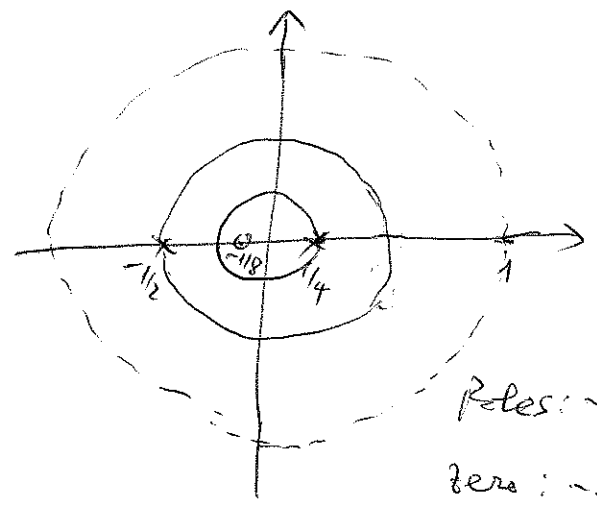


Poles: 1/2, 2
Zero: 1

$$X_3(z) = \frac{2(1 + \frac{1}{8}z^{-1})}{(1 + \frac{1}{2}z^{-1})(1 - \frac{1}{4}z^{-1})}$$

$$X_3(z) = \frac{1}{1 + \frac{1}{2}z^{-1}} + \frac{1}{1 - \frac{1}{4}z^{-1}}$$

- *) $X_3(z)$ is stable for $|z| > \frac{1}{2}$
- *) $X_3(z)$ is causal for $|z| > \frac{1}{2}$
- *) $X_3(z)$ is both stable and causal for $|z| > \frac{1}{2}$



Poles: 1/2, 1
Zero: 1/4

② Let us consider the general case in which $h[n]$ is symmetric around $\frac{n_0}{2}$. Then (similar for anti-symmetric)

$$H(e^{j\omega}) = \sum_{n=-\infty}^{+\infty} h[n] e^{-j\omega n}$$

$$= \sum_{n=-\infty}^{+\infty} h[\underbrace{n_0 - n}_m] e^{-j\omega n}$$

$$= \sum_{m=-\infty}^{+\infty} h[m] e^{-j\omega(n_0 - m)}$$

$$= \sum_{m=-\infty}^{+\infty} h[m] e^{-j\omega n_0} e^{j\omega m}$$

$$= e^{-j\omega n_0} H(e^{-j\omega}) = e^{-j\omega n_0} H^*(e^{j\omega})$$

$$\Rightarrow \angle H(e^{j\omega}) = -\omega n_0 - \angle H(e^{j\omega})$$

$$\Rightarrow \angle H(e^{j\omega}) = -\frac{\omega n_0}{2}$$

$$\Rightarrow \text{grad}(\angle H(e^{j\omega})) = -\frac{d\angle H(e^{j\omega})}{d\omega} = \frac{n_0}{2}$$

Thus:

$$\text{grad}(\angle H_1(e^{j\omega})) = 0.5$$

$$\text{grad}(\angle H_3(e^{j\omega})) = 1.5$$

$$\text{grad}(\angle H_5(e^{j\omega})) = 2$$

$$\text{grad}(\angle H_2(e^{j\omega})) = 2.5$$

$$\text{grad}(\angle H_4(e^{j\omega})) = 3.5$$

$$\text{grad}(\angle H_6(e^{j\omega})) = 0$$

④ From Fig. 2, we see that there are 3 narrow band signals

	ω	Gain	Group delay	Center
Signal 1	0.2π	10	150	$n \approx 80$
Signal 2	0.4π	1.4	0	$n \approx 30$
Signal 3	0.6π	0	-100	$n \approx 150$

↓
Signal 3 disappears after the filter

Thus $y_3[n]$ in Fig. 3 can be the only possible solution